# CIO-DAS6402/12 \& CIO-DAS6402/16 

## Analog and Digital I/O Boards



Revision 2A
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The CIO-DAS6402/16 and CIO-DAS6402/12 provide 32 differential or 64 single-ended inputs. They have sample rates as high as 330 kHz ( 100 kHz for the 16-bit version). Using a 100-pin connector, the CIO-DAS6402 board provides large channel counts without the need for external expansion boards.

The installation and operation of all CIO-DAS6402 series boards is very similar. Throughout this manual we use CIO-DAS6402 as a generic designation for the CIO-DAS6402/12 and CIO-DAS6402/16. When required, due to the differences in the boards, the specific board name is used.

The board's analog input ranges are entirely software-selectable in bipolar ranges from $\pm 10$ to $\pm 1.25$ Volts and unipolar ranges from 0 to 10 V to 0 to 1.25 V . The $\mathrm{A} / \mathrm{D}$ converter can be triggered externally, or internally based on edge or level trigger sources. The board supports standard (post) and pre-trigger operation and so is capable of continuous, scheduled or event-triggered data acquisition.

High speed data acquisition without the chance of a lost sample is assured by the boards' use of the REP INSW data transfers and the on-board 1-Kilosample FIFO memory. In addition to the analog input section, each board also provides two channels of analog output, eight bits of digital input, eight bits of digital output, and a number of counter/timer functions.

All CIO-DAS6402 functions are fully supported by the Universal Library ${ }^{\mathrm{TM}}$ software package. This software provides upper level programming support for all DOS and Windows based operating systems. The CIO-DAS6402 is also shipped with InstaCal ${ }^{\mathrm{TM}}$, a powerful and easy-to-use installation, test, and calibration software package.

We recommend that you install InstaCal ${ }^{\mathrm{TM}}$ before installing the board in your computer. The InstaCal ${ }^{\mathrm{TM}}$ operations will show you how to set the switches and jumpers on the board.

## 2 SOFTWARE INSTALLATION

The CIO-DAS6402 has a variety of switches and jumpers to set before installing the board in your computer. By far the simplest way to configure your board is to use the InstaCal ${ }^{\mathrm{TM}}$ program provided as part of your CIO-DAS6402 software package. InstaCal ${ }^{\mathrm{TM}}$ will show you all available options, how to configure the various switches and jumpers to match your application requirements, and will create a configuration file that your application software (and the Universal Library) will refer to so the software you use will automatically know the exact configuration of the board.

Please refer to the Software Installation Manual regarding the installation and operation of InstaCal ${ }^{\text {TM }}$. The following hard copy information is provided as a matter of completeness, and will allow you to set the hardware configuration of the CIO-DAS6402 board if you do not have immediate access to InstaCal ${ }^{\text {TM }}$ and/or your computer. The locations of each of the switches and jumpers are shown in the figures below.

### 3.1 BASE ADDRESS

Unless there is already a board in your system using address 300h ( 768 Decimal), leave the switches as they are set at the factory. In Figure 3-1, the CIO-DAS6402 is set at base address 300h. See Figures 3-6 and 3-7 for the locations of the switches


BASE ADDRESS SWITCH - Address 300H shown here.

Figure 3-1. Base Address Switches

### 3.2 D/A OUTPUT RANGE SWITCH (CIO-DAS6402/16 ONLY)

The analog output ranges of the CIO-DAS6402/12 are set via software. The analog output ranges of the CIO-DAS6402/16 are set by dip switches on the board (Figure 3-7). Figure 3-2 shows the the allowable switch settings and Table 3-1 gives the range/switch settings.

Figure 3-4. D/A Output Range Switches

Table 3-1. D/A Analog Range Switch Settings

| Range | SW 1 | SW 2 | SW 3 | SW 5 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\pm 10.0 ~ V$ | DN | UP | DN | UP | DN |
| $\pm 5.0 \mathrm{~V}$ | DN | UP | DN | DN | UP |
| $\pm 2.5 \mathrm{~V}$ | DN | UP | DN | DN | DN |
| 0 to 10 V | UP | DN | UP | UP | DN |
| 0 to 5.0 V | UP | DN | UP | UP |  |
| 0 to 2.5 V | UP | DN | UP | DN | DN |

### 3.3 D/A UPDATE MODE (CIO-DAS6402/16 ONLY)

The analog outputs can be configured to update independently or simultaneously. The update mode is set by a jumper on the CIO-DAS6402/16. This jumper is shown in Figure 3-5.

Place the jumper on the XFER side for simultaneous update. Place the jumper on the UPDATE side for independent operation.

Figure 3-5. D/A Update Mode Select Jumper


Figure 3-6. CIO-DAS6402/12 Base Address Switches Location


Figure 3-7. CIO-DAS6402/16 Base Address, D/A Range \& Mode Switches Location

### 3.4 CONNECTOR PIN-OUT

The CIO-DAS6402 analog connector (Figure 3-8) is a 100 pin high-density connector accessible from the rear of the PC through the expansion back plate.

The connector interfaces with the C100FF-2, a two-foot cable. This cable connects directly to the CIO-TERM100 screw terminal board.


CIO-DAS6402 ANALOG SIGNAL CONNECTOR - View from rear of the computer.

Figure 3-8. 100-Pin I/O Connector

### 4.1 ANALOG INPUTS

The following section provides explanations and helpful hints regarding analog input connections. This section is designed to help you achieve the optimum performance from your CIO-DAS6402 series board.

Prior to jumping into actual connection schemes, you should have at least a basic understanding of Single-Ended/Differential inputs and system grounding/isolation. If you are already comfortable with these concepts you may wish to skip to the next section (on wiring configurations).

### 4.1.1 SINGLE-ENDED AND DIFFERENTIAL INPUTS

The CIO-DAS6402 provides either 32 differential or 64 single-ended input channels.

### 4.1.2 SINGLE-ENDED INPUTS

A single-ended input measures the voltage between the input signal and ground. In this case, in single-ended mode the CIO-DAS6402 measures the voltage between the input channel and LLGND. The single-ended input configuration requires only one physical connection (wire) per channel and allows the CIO-DAS6402 to monitor more channels than the (2-wire) differential configuration using the same connector and onboard multiplexor. However, since the CIO-DAS6402 is measuring the input voltage relative to its own low level ground, single-ended inputs are more susceptible to both EMI (Electro Magnetic Interference) and any ground noise at the signal source. Figure $4-1$ shows the theory of single-ended input configuration


## Single-ended input with Common Mode Voltage

Figure 4-1. Single-Ended Voltage Input Theory

## Differential Inputs

Differential inputs measure the voltage between two distinct input signals. Within a certain range (referred to as the common mode range), the measurement is almost independent of signal source to CIO-DAS6402 ground variations. A differential input is also much more immune to EMI than a single-ended one. Most EMI noise induced in one lead is also induced in the other, the input only measures the difference between the two leads, and the EMI common to both is ignored. This effect is a major reason there is twisted pair wire as the twisting assures that both wires are subject to virtually identical external influence. Figure 4-2 below shows a typical differential input configuration.


Figure 4-2. Differential input Theory
Before moving on to the discussion of grounding and isolation, it is important to explain the concepts of common mode, and common mode range (CM Range). Common mode voltage is depicted in the diagram above as Vcm. Though differential inputs measure the voltage between two signals, without (almost) respect to the either signal's voltages relative to ground, there is a limit to how far away from ground either signal can go. Though the CIO-DAS6402 has differential inputs, it will not measure the difference between 100 V and 101 V as 1 Volt (in fact the 100 V would destroy the board!). This limitation or common mode range is depicted graphically in Figure 4-3. The CIO-DAS6402 common mode range is $+/-10$ Volts. Even in differential mode, no input signal can be measured if it is more than 10V from the board's low level ground (LLGND).


Figure 4-3. Common Mode Range

### 4.1.3 SYSTEM GROUNDS AND ISOLATION

There are three scenarios possible when connecting your signal source to your CIO-DAS6402 board.

1. The CIO-DAS6402 and the signal source may have the same (or common) ground. This signal source may be connected directly to the CIO-DAS6402.
2. The CIO-DAS6402 and the signal source may have an offset voltage between their grounds (AC and/or DC). This offset it commonly referred to a common mode voltage. Depending on the magnitude of this voltage, it may or may not be possible to connect the CIO-DAS6402 directly to your signal source. We will discuss this topic further in a later section.
3. The CIO-DAS6402 and the signal source may already have isolated grounds. This signal source may be connected directly to the CIO-DAS6402.

## Which system do you have?

Try the following test: Using a battery powered voltmeter*, measure the voltage between the ground signal at your signal source and at your PC. Place one voltmeter probe on the PC ground and the other on the signal source ground. Measure both the AC and DC Voltages.
*If you do not have access to a voltmeter, skip the experiment and read the following three sections. You may be able to identify your system type from the descriptions provided.

If both AC and DC readings are 0.00 volts, you may have a system with common grounds. However, since voltmeters will average out high frequency signals, there is no guarantee. Please refer to the section below titled Common Grounds.

If you measure reasonably stable AC and DC voltages, your system has an offset voltage between the grounds category. This offset is referred to as a Common Mode Voltage. Please be careful to read the following warning and then proceed to the section describing Common Mode systems.

## WARNING

If either the AC or DC voltage is greater than 10 volts, do not connect the CIO-DAS6402 to this signal source. You are beyond the boards usable common mode range and will need to either adjust your grounding system or add special Isolation signal conditioning to take useful measurements. A ground offset voltage of more than 30 volts will likely damage the CIO-DAS6402 board and possibly your computer. Note that an offset voltage much greater than 30 volts will not only damage your electronics, but it may also be hazardous to your health.

This is such an important point, that we will state it again. If the voltage between the ground of your signal source and your PC is greater than 10 volts, your board will not take useful measurements. If this voltage is greater than 30 volts, it will likely cause damage, and may represent a serious shock hazard! In this case you will need to either reconfigure your system to reduce the ground differentials, or purchase and install special electrical isolation signal conditioning.

If you cannot obtain a stable DC voltage measurement between the grounds, or the voltage drifts around considerably, the two grounds are most likely isolated. The easiest way to check for isolation is to change your voltmeter to it's ohm scale and measure the resistance between the two grounds. It is recommended that you turn both systems off prior to taking this resistance measurement. If the measured resistance is more than 100 Kohm , it's a fairly safe bet that your system has electrically isolated grounds.

## Systems with Common Grounds

In the simplest (but perhaps least likely) case, your signal source will have the same ground as the CIO-DAS6402. This would typically occur when providing power or excitation to your signal source directly from the CIO-DAS6402. There may be other common ground configurations, but it is important to note that any voltage between the CIO-DAS6402 ground and your signal ground is a potential error voltage if you set up your system based on a common ground assumption.

As a safe rule of thumb, if your signal source or sensor is not connected directly to an LLGND pin on your CIO-DAS6402, it's best to assume that you do not have a common ground even if your voltmeter measured 0.0 Volts. Configure your system as if there is ground offset voltage between the source and the CIO-DAS6402. This is especially true if you are using either the CIO-DAS6402/16 or the CIO-DAS6402/12 at high gains, since ground potentials in the sub millivolt range will be large enough to cause A/D errors, yet will not likely be measured by your handheld voltmeter.

## Systems with Common Mode (ground offset) Voltages

The most frequently encountered grounding scenario involves grounds that are somehow connected, but have AC and/or DC offset voltages between the CIO-DAS6402 and signal source grounds. This offset voltage may be AC, DC or both and may be caused by EMI pickup, resistive voltage drops in ground wiring and connections, etc. Ground offset voltage is a more appropriate term to describe this type of system, but since our goal is to keep things simple, and help you make appropriate connections, we'll stick with our somewhat loose usage of the phrase Common Mode.

## Small Common Mode Voltages

If the voltage between the signal source ground and CIO-DAS6402 ground is small, the combination of the ground voltage and input signal will not exceed the CIO-DAS6402's $+/-10 \mathrm{~V}$ common mode range, (i.e. the voltage between grounds, added to the maximum input voltage, stays within $+/-10 \mathrm{~V}$ ), This input is compatible with the CIO-DAS6402 and the system may be connected without additional signal conditioning. Fortunately, most systems will fall in this category and have a small voltage differential between grounds.

## Large Common Mode Voltages

If the ground differential is large enough, the CIO-DAS6402's $+/-10 \mathrm{~V}$ common mode range will be exceeded (i.e. the voltage between CIO-DAS6402 and signal source grounds, added to the maximum input voltage you're trying to measure exceeds $+/-10 \mathrm{~V}$ ). In this case the CIO-DAS6402 cannot be directly connected to the signal source. You will need to change your system grounding configuration or add isolation signal conditioning. (Please look at our ISO-RACK and ISO-5B-series products to add electrical isolation, or give our technical support group a call to discuss other options.)

NOTE
Do not rely on the earth prong of a 120 VAC receptacle for signal ground connections. Different ground plugs may have large and potentially even dangerous voltage differentials. Remember that the ground pins on 120VAC outlets on different sides of the room may only be connected in the basement. This leaves the possibility that the "ground" pins may have a significant voltage differential (especially if the two 120VAC outlets happen to be on different phases.)

## CIO-DAS6402 and signal source already have isolated grounds

Some signal sources will already be electrically isolated from the CIO-DAS6402. The diagram below shows a typical isolated ground system. These signal sources are often battery powered, or are fairly expensive pieces of equipment (since isolation is not an inexpensive proposition), isolated ground systems provide excellent performance, but require some extra effort during connections to assure optimum performance is obtained. Please refer to the following sections for further details.

### 4.2 WIRING CONFIGURATIONS

Combining all the grounding and input type possibilities provides us with the following potential connection configurations. The combinations along with our recommendations on usage are shown in Table 4-1 below.

Table 4-1. Input vs. Grounding Recommendations

| Ground Category | Input Configuration | Our view |
| :---: | :---: | :---: |
| Common Ground | Single-Ended Inputs | Recommended |
| Common Ground | Differential Inputs | Acceptable |
| Common Mode <br> Voltage < +/-10V | Single-Ended Inputs | Not Recommended |
| Common Mode <br> Voltage < +/-10V | Differential Inputs | Recommended |
| Common Mode <br> Voltage > +/- 10V | Single-Ended Inputs | Unacceptable without adding Isolation |
| Common Mode <br> Voltage > +/-10V | Differential Inputs | Unacceptable without adding Isolation |
| Already Isolated Grounds | Single-ended Inputs | Acceptable |
| Already Isolated Grounds | Differential Inputs | Recommended |

The following sections depicts recommended input wiring schemes for each of the eight possible input configuration/grounding combinations.

### 4.2.1 COMMON GROUND / SINGLE-ENDED INPUTS

Single-ended is the recommended configuration for common ground connections. However, if some of your inputs are common ground and some are not, we recommend you use the differential mode. There is no performance penalty (other than loss of channels) for using a differential input to measure a common ground signal source. However the reverse is not true. Figure 4-4 below shows a recommended connection diagram for a common ground / single-ended input system


Signal source and A/D board sharing common ground connected to single-ended input.

Figure 4-4. Common Ground / Single-Ended Inputs

### 4.2.2 COMMON GROUND / DIFFERENTIAL INPUTS

The use of differential inputs to monitor a signal source with a common ground is a acceptable configuration though it requires more wiring and offers fewer channels than selecting a single-ended configuration. Figure $4-5$ below shows the recommended connections in this configuration.


> Signal source and A/D board sharing common ground connected to differential input.

Figure 4-5. Common Ground / Differential Inputs

### 4.2.3 COMMON MODE VOLTAGE < +/-10V / SINGLE-ENDED INPUTS

This is not a recommended configuration. In fact, the phrase common mode has no meaning in a single-ended system and this case would be better described as a system with offset grounds. Anyway, you are welcome to try this configuration, no system damage should occur and depending on the overall accuracy you require, you may receive acceptable results.

### 4.2.4 COMMON MODE VOLTAGE < +/-10V / DIFFERENTIAL INPUTS

 input signal must stay within $+/-10 \mathrm{~V}$

> Signal source and A/D board with common mode voltage connected to a differential input.

Systems with varying ground potentials should always be monitored in the differential mode. Care is required to assure that the sum of the input signal and the ground differential (referred to as the common mode voltage) does not exceed the common mode range of the A/D board ( $+/-10 \mathrm{~V}$ on the CIO-DAS6402). Figure $4-6$ below show recommended connections in this configuration.

Figure 4-6. Common Mode Voltage < + /-10V / Differential Inputs

### 4.2.5 COMMON MODE VOLTAGE > +/-10V

The CIO-DAS6402 will not directly monitor signals with common mode voltages greater than $+/-10 \mathrm{~V}$. You will either need to alter the system ground configuration to reduce the overall common mode voltage, or add isolated signal conditioning between the source and your board. See Figure 4-7 and 4-8 below.


Figure 4-7. Common Mode Voltage $>+/-10 V$. Single-Ended Input


> When the voltage difference between signal source and A/D board ground is large enough so the $A / D$ board's common mode range is
exceeded, isolated signal
10K is a recommended value. You may short LL GND to CH Low
10K is a recommended value. You may short LL
instead, but this will reduce your system's noise immunity.

System with a Large Common Mode Voltage,
Connected to a Differential Input

Figure 4-8. Common Mode Voltage > +/-10V. Differential Input

### 4.2.6 ISOLATED GROUNDS / SINGLE-ENDED INPUTS

Single-ended inputs can be used to monitor isolated inputs, though the use of the differential mode will increase you system's noise immunity. Figure $4-9$ below shows the recommended connections is this configuration.


Isolated Signal Source
Connected to a Single-Ended Input

Figure 4-9. Isolated Grounds / Single-Ended Input

### 4.2.7 ISOLATED GROUNDS / DIFFERENTIAL INPUTS

Optimum performance with isolated signal sources is assured with the use of the differential input setting. Figure 4-10 below shows the recommend connections is this configuration.


Already isolated signal source and $A / D$ board connected to a differential input.

Figure 4-10. Isolated Grounds / Differential Inputs

### 4.3 ANALOG OUTPUTS

Analog outputs are simple voltage outputs which can be connected to any device which will record, display or be controlled by a voltage. The CIO-DAS6402 analog outputs are 4 quadrant multiplying DACs. This means that they accept an input voltage reference and provide an output voltage which is inverse to the reference voltage and proportional to the digital value in the output register.

For example, the supplied reference of -5 V provides a +5 V output when the value in the output register is 4095 (full scale at 12 -bits of resolution). It provides a value of 2.5 V when the value in the output register is 2048 .

The output ranges of the CIO-DAS6402/12 are set by software. The output ranges of the CIO-DAS6402/16 are set by dip switch. Please refer to Section 3 for information regarding setting these switches.

### 5.1 INTRODUCTION

The CIO-DAS6402 is controlled and monitored by reading and writing to 16 I/O addresses. The first address is referred to as the BASE ADDRESS and is set by a bank of switches on the board. All other addresses are located at the BASE ADDRESS plus a specified offset.

Registers are easy to read from and write to, though to create a complete data acquisition software program at the register level is a significant undertaking. Unless there is a specific reason that you need to write your program at the register lever, we recommend using our Universal Library.

In summary form, the registers and their function are listed on the following table. Within each register are eight bits which may constitute a byte of data or be eight individual bit set/read functions.

Table 5-1. DAS6402 I/O Map - COMPATIBLE Mode
(BOLD indicates register definition for DAS6402/16)

| ADDRESS | READ FUNCTION | WRITE FUNCTION |
| :--- | :--- | :--- |
| BASE | A/D bits 0(LSB) - 3 \& Channel \# | Software Start A/D Conversion |
| BASE +1 | A/D Bits $4-11(\mathrm{MSB})$ | None |
| BASE | A/D bits 0 (LSB) - 7 | Software Start A/D Conversion |
| BASE $+\mathbf{1}$ | A/D Bits 8 - 15 (MSB) | None |
| BASE +2 | $8 / 16$ Channel Mux | $8 / 16$ Channel Mux / Reset FIFO |
| BASE +3 | Digital Input bits 0-3 / External control | Digital Output Bits 0-3 |
| BASE +4 | None | D/A 0 Bits 0-3 |
| BASE +5 | None | D/A 0 Bits 4-11 |
| BASE +6 | None | D/A 1 Bits 0-3 |
| BASE +7 | None | D/A 1 Bits 4-11 |
| BASE +8 | EOC, UNI/BIP, SEDIFF, Mux | Reset interrupt flip-flop |
| BASE +9 | DMA Enable, Interrupt, Pacer Source | DMA Enable, Interrupt, Pacer Source |
| BASE +10 | DAC range, TRIG0/CTR0 | DAC range, TRIG0/CTR0 |
| BASE +11 | DMA, U/B, SEDIFF, MODE, A/D Gain | MODE, A/D Gain Control |
| BASE +12 | Counter 0 Data | Counter 0 Data |
| BASE +13 | CTR 1 Data - A/D Pacer Clock | CTR 1 Data - A/D Pacer Clock |
| BASE +14 | CTR 2 Data - A/D Pacer Clock | CTR 2 Data - A/D Pacer Clock |
| BASE +15 | None. No read back on 8254 | 8254 Counter Control Register |

Table 5-2. DAS6402 I/O Map - ENHANCED Mode
(BOLD indicates register definition for DAS6402/16)

| ADDRESS | READ FUNCTION | WRITE FUNCTION |
| :--- | :--- | :--- |
| BASE | A/D bits 0(LSB) -11 (MSB) (Word) | Software Start A/D Conversion |
| BASE | A/D bits 0(LSB) -15 (MSB) (Word) | Software Start A/D Conversion |
| BASE +1 | Do not use, use BASE only | None |
| BASE +2 | Post-Trigger Index Counter | $32 / 64$ Channel Mux (Word) /Reset FIFO |
| BASE +3 | Digital Input bits 0-7 / External control | Digital Output Bits 0-7 |
| BASE +4 | None | D/A 0 Bits 0-3 |
| BASE +5 | None | D/A 0 Bits 4-11 |
| BASE +6 | None | D/A 1 Bits 0-3 |
| BASE +7 | None | D/A 1 Bits 4-11 |
| BASE +8 | FIFO, Interrupt Status, Clock rate | Extend modes and clear interrupts |
| BASE +9 | Burst Enable, Interrupt, Pacer Source | Burst Enable, Interrupt, Pacer Source |
| BASE +10 | DAC range, Pacer clock control | DAC range, Pacer clock control |
| BASE +11 | UB,SEDIFF,MODE,Int source, ADGain | UB,SEDIFF,MODE,Int source, ADGain |
| BASE +12 | CTR 0 Data: (Residual/ End of acq) | CTR 0 Data: Preload for residual count |
| BASE +13 | CTR 1 Data - A/D Pacer Clock | CTR 1 Data - A/D Pacer Clock |
| BASE +14 | CTR 2 Data - A/D Pacer Clock | CTR 2 Data - A/D Pacer Clock |
| BASE +15 | None. No read back on 8254 | 8254 Counter Control Register |

### 5.2 A/D DATA WORD REGISTERS

In Enhanced Mode, two of the data registers are configured for 'Word' reads or writes, as opposed to performing the I/O operation as separate byte reads or writes. These two registers are the A/D Data Register at the BASE +0 , and the Channel Mux $\mathrm{Hi} / \mathrm{Lo}$ Register at BASE+2. Any IO accesses (reads or writes) to either of these registers is interpreted by the board as Word IO.

## A/D DATA WORD REGISTER - 12 BIT

BASE + 0 Example, 300h, 768 Decimal
READ/WRITE

| Mode | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENHANCED | AD11 <br> MSB | AD10 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 <br> LSB | 0 | 0 | 0 | 0 |
| COMPATIBLE | AD11 <br> MSB | AD10 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 <br> LSB | MA3 | MA2 | MA1 | MA0 |

A read/write register. The A/D Data Register is configured as a word because REP INSW can be used to quickly read data from the board, allowing for higher A/D conversion rates than would be possible if using DMA, which accesses the A/D data as two bytes.

READ
On read, the 12-bit ADC value is presented in 'left-justified' format, with the most-significant ADC bit occupying the data word bit position \#15; the least-significant ADC bit occupies bit position \#4 of the data word.

WRITE
A write to the base address will cause an A/D conversion, (Bits $0 \& 1$ of BASE +9 must be 0 .)

ENHANCED MODE: The channel tag is not available in enhanced mode, thus the lowest four bits of the data word will be read back as zero.

COMPATIBLE MODE: The channel tag is available in compatible mode, occupying the lowest four bits of the data word.

## A/D DATA WORD REGISTER - 16 BIT

BASE $+0 \quad$ Example, 300h, 768 Decimal
READ/WRITE

| Mode | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENHANCED | AD15 <br> MSB | AD14 | AD13 | AD12 | AD11 | AD10 | AD9 9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 <br> LSB |
| COMPATIBLE | AD15 <br> MSB | AD14 | AD13 | AD12 | AD11 | AD10 | AD9 | AD8 | AD7 | AD6 | AD5 | AD4 | AD3 | AD2 | AD1 | AD0 <br> LSB |

## READ

On read, the 16-bit ADC value is presented in 'left-justified' format, with the most-significant ADC bit occupying the data word bit position \#15; the least-significant ADC bit occupies bit position \#0 of the data word.

WRITE
A write to the base address will cause an A/D conversion, (Bits $0 \& 1$ of BASE +9 must be 0 .)
Whether in ENHANCED or COMPATIBLE MODE, there is no channel tag.

### 5.3 CHANNEL MUX HI/LO LIMITS WORD REGISTER

BASE ADDRESS +2 Example, 302h, 770 Decimal WRITE

| Mode | 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENHANCED | - | - | HI 5 | HI 4 | HI 3 | HI 2 | HI 1 | HI 0 | - | - | LO5 | LO4 | LO3 | LO2 | LO1 | LO0 |
| COMPATIBLE | - | - | - | - | - | - | - | - | 8 H | 4 H | 2 H | 1 H | 8 L | 4 L | 2 L | 1 L |

## WRITE

Writing to this register clears the FIFO.

## ENHANCED MODE:

This register is configured as a Word to allow for additional bits, two for the low channel limit and two for the high channel limit, which are used to generate the 6 -bit counter sequence. This register cannot be accessed as two separate byte writes, access must be done as an entire word. If a byte write to address 302 is attempted, then the HI0-HI5 data will be undefined. If a byte write to address 303 is attempted, the board will assume that the data is to be written to the DOUT register.
When this register is written, the analog input multiplexers are set to the channel specified in LO0-LO5. After each conversion, the input multiplexers increment to the next channel, reloading to the "LO" channel after the " HI " channel is reached.

## READ

Used to return the pre-trigger index value as was done by an 8254 counter.

## COMPATIBLE MODE:

READ/WRITE
The mux register operates as the DAS1600 series. The lower eight bits are separated in two nibbles; the upper four bits represent the upper channel and the lower four bits represent the lower channel. For example: to scan channels 2 through 5, write the value 52 to Base +2 . The upper eight bits in this register are written to but cannot be read, and do not affect the mux sequencing circuit.

### 5.4 8-BIT DIGITAL I/O REGISTERS

BASE ADDRESS + 3
Example, 303h, 771 Decimal
READ

| 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DI7 | DI6 | DI5 | DI4 | DI3 | DI2 | DI1 | DI0 |
|  |  |  |  | GATE0 | XTRIG | XPACER |  |

The signals present at the 8 digital inputs are read as one byte. Three of the pins have special functions in addition to digital input.:

| XPACER/DI0 | External Pacer: | Starts an A/D Conversion on each active edge. |
| :--- | :--- | :--- |
| XTRIG/DI1 | External Trigger: | Causes an entire acquisition to start or stop. |
| GATE0/DI2 | Gate for CTR0 | Used in Compatible mode only. |

These special functions are optionally enabled in software, and are explained below. External Interrupt is enabled at BASE +9 , External Pacer and External Trigger are enabled at BASE +10 . Please see those register descriptions for details.

WRITE

| 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DO7 | DO6 | DO5 | DO4 | DO3 | DO2 | DO1 | DO0 |

All of the eight bits are latched TTL outputs.

### 5.5 DIGITAL TO ANALOG CONVERTER (ANALOG OUT) REGISTERS

## D/A 0 REGISTERS

BASE ADDRESS +4 Example, 304h, 772 Decimal

| 6402/ | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12-BIT | D/A3 | D/A2 | D/A1 | D/A0 | 0 | 0 | 0 | 0 |
| 16-BIT | D/A7 | D/A6 | D/A5 | D/A4 | D/A3 | D/A2 | D/A1 | D/A0 |

BASE ADDRESS +5 Example, 305h, 773 Decimal

| 6402/ | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12-BIT | D/A11 | D/A10 | D/A9 | D/A8 | D/A7 | D/A6 | D/A5 | D/A4 |
| 16-BIT | D/A15 | D/A14 | D/A13 | D/A12 | D/A11 | D/A10 | D/A9 | D/A8 |

## D/A 1 REGISTERS

BASE ADDRESS +6 Example, 306h, 774 Decimal

| 6402/ | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12-BIT | D/A3 | D/A2 | D/A1 | D/A0 | 0 | 0 | 0 | 0 |
| 16-BIT | D/A7 | D/A6 | D/A5 | D/A4 | D/A3 | D/A2 | D/A1 | D/A0 |

BASE ADDRESS +7 Example, 307h, 775 Decimal

| 6402/ | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12-BIT | D/A11 | D/A10 | D/A9 | D/A8 | D/A7 | D/A6 | D/A5 | D/A4 |
| 16-BIT | D/A15 | D/A14 | D/A13 | D/A12 | D/A11 | D/A10 | D/A9 | D/A8 |

WRITE ONLY

## 6402/12

Each 12 bit D/A output line has two 8-bit registers. The first contains the four least-significant bits of the data and four 'don't-care' bits. The second register contains the eight most significant bits of the data.

Data can be written as two successive bytes or as one word. When two bytes are written, the lower address byte must be written first, and then the higher byte, as the DAC output is updated when the higher byte is written.

When the data is written as a single word, the most significant bit of the DAC data is the most significant bit of the word. The CPU will automatically write the data to the board as two successive bytes, writing the low and then the high byte.

## 6402/16

Each 16-bit D/A output line has two 8 -bit registers. The first contains the eight least significant bits of the data and the second register contains the eight most significant bits of the data. The data is written as two successive bytes - the lower first (or LSB's) followed by the higher byte (or MSB's).
The 6402/16 also has a jumper (HD1) to select either individual update (UPDATE) or simultaneous updating (XFER) of both DAC's. In UPDATE mode, when the higher address is written, that DAC is updated. In XFER mode, after both DAC's are loaded, a Read cycle to any of the DAC registers with update both DAC's simultaneously.

### 5.6 STATUS REGISTER

BASE ADDRESS +8 Example, 308h, 776 Decimal
READ

| Mode | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENHANCED | $1 / 10 \mathrm{MHz}$ | INDGT | XTRIG | INT | XINT | FFULL | FHALF | FFNE |
| COMPATIBLE | EOC | UNIBIP | SEDIFF | INT | MA3 | MA2 | MA1 | MA0 |

Description of Status Register read bits:

## ENHANCED MODE:

$\mathbf{1 / 1 0 M H z}$ - Internal Pacer clock source. $=1$, Pacer clock is $10 \mathrm{MHz},=0$, Pacer clock is 1 MHz .

INDGT - Index Gate. Signals when the index counter has been shut off due to FIFO going Half Full. When = 0, index counter active or flip-flop has been reset (by CLRXTR). When $=1$, index counter has been shut off.

XTRIG - State of external trigger flop, used when edge-triggering. When $=1$, the trigger has been activated. When $=0$ idle or not active. Cleared by CLRXTR.

INT - State of interrupt flop, from 1 of 4 sources on the board. When $=1$, the flip-flop indicates an interrupt condition. When $=0$ no interrupt. Cleared by CLRINT. See BASE +9 for interrupt enable and interrupt level and BASE +11 for interrupt sources.

XINT - State of external interrupt flop, acts independently of on-board interrupt sources. When $=1$, external interrupt occurred, When $=0$ no external interrupt. Cleared by CLRXIN.

FFULL - latched status of FIFO Full condition.
When $=1$, FIFO exceeded full state, data may have been lost. When $=0$, FIFO has not exceeded full state.
FHALF - indicates if FIFO is above or below half-full. Can be used as an interrupt source for REP INSW ADC.
When $=1$, FIFO is above half-full. When $=0$, FIFO is at or below half-full. (not latched)
FFNE - Not empty state of FIFO. Can be used as interrupt for single conversion ADC mode.
When $=1$, FIFO is not empty (contains ADC data). When $=0$, FIFO is empty (has no ADC data).

## COMPATIBLE MODE:

EOC - End-of-conversion, When $1=$ busy. When $=0$, conversion complete .
UNIBIP - Analog input Unipolar/Bipolar status, When $1=$ unipolar. When 0 , = bipolar.

SEDIFF - Analog input channel configuration, $0=$ differential, $1=$ single ended.
MA3:0 - Analog input channel mux setting (for next conversion).
WRITE

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{1 / 1 0 M H z}$ | POSTMODE | ARMED | EXTEND | - | CLRXIN | CLRXTR | CLRINT |

## ENHANCED MODE:

The write functions of the status register are to clear certain flip-flop states, and set the Pacer clock to 1 or 10 MHz .
CLRINT - Clear Interrupt flip-flop when $=1$. No action when $=0$.
CLRXTR - Clear External Trigger flip-flop when $=1$. No action when $=0$
CLRXIN - Clear External Interrupt flip-flop when $=1$. No action when $=0$.
(These 'Clear' functions get automatically reset to 0 in 1 to $2 \mu \mathrm{sec}$ ).
EXTEND - Qualifier for bits 5 to 7 data. Must be set to 1 prior to setting the desired bits at 5 to 7 .
$=0$, any data at bits 5 to 7 will be ignored and the state of the bit will not change.
The MODE bit must be $=1$ (Enhanced) to change the EXTEND bit.
Example: to set the $1 / 10 \mathrm{MHz}$ bit to set the Pacer time base to 10 MHz :
Do each step as separate write operations:

1. Set MODE bit to 1 (base $+B$, bit 4)
2. Set EXTEND to 1 (base +8 , bit 4 )
3. Set $1 / 10$ bit to 1 (base +8 , bit 7 ), be sure to leave bit 4 set as well, including bits $5 / 6$ if desired to leave them set.
4. Set EXTEND bit to 0, leave the bits in 5 to 7 set as required.

After EXTEND bit is zeroed, any writes to bits 5 to 7 are masked off (doesn't change).
Please note that the clear flip-flop bits in 0 to 2 will be active if their corresponding data bits are high during any write to base +8 in Enhanced Mode.

ARMED - Used with various pre and post-trigger scenarios to gate a 'residual' counter on (CTR0 of the 8254).
Counter 0 operates in Mode 0 and counts conversions, stopping the acquisition process when it reaches terminal count.
When $=0$, counting of post-trigger samples is disabled. When $=1$, counting of post-trigger samples is enabled.
See also POSTMODE.

POSTMODE - Used to select whether FifoHalf Full or ADC convert starts Counter 0. Used in conjunction with PRETRIG to select a strobe to latch the Counter 0 gate. (See BASE +10 for PRETRIG.) Refer to Table 6-3.

Total \# of counts (N) less- than-or-equal-to FIFO 1/2-full is a special case:
Table 5-3. Post-Mode Operations

| Gate CTR0: | Pretrig? | POST-MODE | Ctr 0 gate strobe: | ARMED bit is set by: |  |
| :---: | :---: | :---: | :---: | :---: | :--- |
| fifo $>\mathrm{N}>1 / 2$ fifo | no | $\mathbf{0}$ | $\mathbf{0}$ | on last fifo half full | Before acquisition is started |
| $\mathrm{N}>$ fifo | no | $\mathbf{0}$ | $\mathbf{0}$ | on last fifo half full | Next to last full 'packet' |
| $\mathrm{N}<=1 / 2$ fifo | no | $\mathbf{0}$ | $\mathbf{1}$ | 1st conversion | Before acquisition is started |
| $\mathrm{N}>1 / 2$ fifo | yes $\mathbf{1}$ | $\mathbf{0}$ | on fifo half (last interrupt $)$ | ISR on last full 'packet' |  |
| $\mathrm{N}<=1 / 2$ fifo | yes $\mathbf{1}$ | $\mathbf{1}$ | on the external trigger | Before acquisition is started |  |

$\mathbf{1 / 1 0 M H z}$ - Sets Pacer clock frequency. .
When bit $7=1$, Forces Pacer clock to 10 MHz . When $=0$, forces Pacer clock to 1 MHz .

Write to Base +8 only clears the interrupt flip-flop (sets CLRINT only (=1)).

### 5.7 INTERRUPT AND PACER CONTROL REGISTER

BASE ADDRESS +9 Example, 309h, 777 Decimal

## READ/WRITE

|  | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENH | INTE | HC_IS2 | HC_IS1 | HC_IS0 | XINTE | BURSTE | HC_PS1 | HC_PS0 |
| COM | INTE | HC_IS2 | HC_IS1 | HC_IS0 | XINTE | DMAEN | HC_PS1 | HC_PS0 |

INTE $=1$, Analog Interrupts are enabled, $=0$, interrupts are disabled.
The system interrupt level is determined by the three HC_ISO-2 bits. The interrupt source (Single AD, FIFO Half-Full, End of Burst) is selected by the two Analog Interrupt Source bits HC_AI0 \& HC_AI1 (Base + 11). Use of external interrupt (XINTE) also requires that INTE $=1$.

HC_IS2, HC_IS1, HC_IS0 will map an enabled interrupt onto a specific ISA bus interrupt level. See Table 6-4 below:
Table 5-4. Interrupt Levels - Enhanced \& Compatible

| HC_IS2 | HC_IS1 | HC_IS0 | Interrupt Level- Enhanced | Compatible (ref) |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | none | none |
| 0 | 0 | 1 | 11 | none |
| 0 | 1 | 0 | 2 | 2 |
| 0 | 1 | 1 | 3 | 3 |
| 1 | 0 | 0 | 10 | none - see Note |
| 1 | 0 | 1 | 5 | 5 |
| 1 | 1 | 0 | 15 | 6 |
| 1 | 1 | 1 | 7 | 7 |

Note: Some Interrupt levels in Enhanced Mode are different from Interrupt levels in Compatible Mode. IRQ4 was dropped from the Compatible-mode IRQ's to allow a single GAL to decode all IRQ's, and since COM1 uses IRQ4, the DAS1600 will probably never use it.

XINTE: $=1$, External Interrupt enabled. $=0$, External Interrupt disabled. The external interrupt is a direct digital path from the 100-pin connector to the system interrupt. External Interrupt shares the Interrupt Level with the Analog Interrupt path. (Only enabled in Enhanced mode).
BURSTE When $=1$, Burst Mode is enabled. When $=0$, Burst Mode is disabled. The number of channels in the burst is determined by the Mux Hi/Lo register.
(BURST Mode also enabled by write to BASE +406 hex bit 6 as on DAS1600 series).
HC_PS1, HC_PS0: Controls the source of the A/D conversion Pacer according Table 6-5 below:
Table 5-5. Control Source of A/D Conversion Pacer

| HC_PS1 | HC_PS0 | Pacer Source |
| :---: | :---: | :---: |
| 0 | 0 | Software Convert |
| 0 | 1 | External Pacer Falling Edge |
| 1 | 0 | External Pacer Rising Edge |
| 1 | 1 | Internal Pacer (8254) |

### 5.8 TRIGGER CONTROL/ DAC RANGE SELECT REGISTER

Triggering and Gating are digital means to control pacing. Triggering means that an active edge on the DI1 pin will start the A/D Pacer. When the active edge occurs, the state of the pin is Don't Care. Gating means that a digital level on the DI1 pin will start or stop the A/D Pacer; when the gate is active, A/D conversions are enabled. When the gate is inactive, A/D conversions are suspended but they will resume when the gate returns to its active state. Gating and triggering both share DI1 pin and thus are mutually exclusive. External Pacing, where each A/D conversion is started by an external signal, is tied to the DI0 pin. This means that External Pacing can be combined with a Trigger/Gate function.

BASE ADDRESS +10 Example, 30Ah, 778 Decimal
WRITE

| Mode | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENHANCED | DAC1R1 | DAC1R0 | DAC0R1 | DAC0R0 | PRETRIG | T/G POL | T/G SEL | T/G EN |
| COMPATIBLE | DAC1R1 | DAC1R0 | DAC0R1 | DAC0R0 | - | - | CTR0 | TRG0 |

NOTE: DACxRx ARE USED WITH THE 6402/12 ONLY, (FOR THE 6402/16 THESE BITS HAVE NO FUNCTION - THE DAC'S RANGES ARE SELECTED BY SWITCHES)

DAC\#R0: Output Range select bit for DAC \#. When $=1$, range is $\pm 5 \mathrm{~V}$ or 0 to 5 V . When $=0$; range is $\pm 10 \mathrm{~V}$ or 0 to 10 V . DAC\#R1: Output Polarity select bit for DAC \#. When $=1$, polarity is Unipolar. When $=0$, polarity is Bipolar

Table 5-6. Output Range Select Codes

| Set DAC Output Range | DAC\#R1 | DAC\#R0 |
| :---: | :---: | :---: |
| $\pm 5 \mathrm{~V}$ | 0 | 0 |
| $\pm 10 \mathrm{~V}$ | 0 | 1 |
| 0 to 5 V | 1 | 0 |
| 0 to 10 V | 1 | 1 |

## ENHANCED MODE:

T/G EN: Trigger/Gate Enable bit
When $=1$, DI1 pin is enabled for Trigger/Gate function.
When $=0$, Trigger/Gate function is not used. DI1 pin is strictly used for digital input.
If pretriggering, then set T/G EN to 0 since this will gate in the CTR0OUT clock with ARMED.
T/G SEL: $\quad$ Trigger/Gate Select bit (Requires T/G EN =1)
When $=0$, DI1 pin functions as a Gate, the active level is determined by the T/G POL bit.
When $=1$, DI1 pins functions as a Trigger, the active edge is determined by T/G POL bit.
T/G POL: $\quad$ Trigger/Gate Polarity bit (Requires T/G EN =1 or PRETRIG = 1)
When $=0$, if trigger, active edge is rising; if gate, active level is high.
When $=1$, if trigger, active edge is falling; if gate, active level is low.
PRETRIG: Used to stop pacing a certain number of conversions after the trigger occurs.
When $=1$, Pre-trigger Mode enabled. $=0$, Pre-trigger Mode disabled.
8254 Counter 0 is used as the 'Post-Trigger' counter
(Pre-trigger mode requires that $\mathrm{T} / \mathrm{G} \mathrm{EN}=0$, and that $\mathrm{T} / \mathrm{G} \mathrm{SEL}=1$ for edge triggering)
Counter 0 clock input is generated from the Conversion Complete pulse to count the number of conversions performed.
A 'Pre-trigger index' counter value can be read (as a Word) from Base +2 to give the number of conversions that have occurred from the Trigger to the half-full or End-of-Acquisition interrupt. Since current FIFO half-full is 512 conversions this number will be in the range of 0 to 511 .

## COMPATIBLE MODE:

TRIG0: Trigger/Gate Enable bit
When $=1$, Enables DIN0 input to gate the pacer (external trigger/gate).
When $=0$, Enables pacer gate preventing external trigger/gate control.
CTR0: $\quad$ Counter 0 input control
When $=1,100 \mathrm{khz}$ input to counter 0 if external counter 0 input is pulled high (or unconnected since the pin is internally pulled high).
When $=0$, external clock input generates pulses to counter 0 clock input.
READ

| Mode | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ENHANCED | - | POSTMODE | ARMED | PACERGATE | PRETRIG | T/G POL | T/G SEL | T/G EN |
| COMPATIBLE | - | POSTMODE | ARMED | PACERGATE | - | - | CTR0 | TRG0 |

ARMED - When $=0$, counting of post-trigger samples is disabled. When $=1$, counting of post-trigger samples is enabled.

## POSTMODE - SEE REGISTER BASE+8 FOR FULL DESCRIPTION.

PACERGATE - When $=0$ : the Gate for the 8254 PACER is off. When $=1$, the Pacer is on.

### 5.9 COMPATIBLE MODE CONTROL REGISTER

BASE ADDRESS +11 Example, 30Bh, 779 Decimal

## READ/WRITE

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DMA1/3 | UNI/BIP | SE/DIFF | MODE | HC_AI1 | HC_AI0 | GAIN1 | GAIN0 |

Since the DAS6402 has no switches or jumpers, certain options must be selectible for Compatible Mode. This is done by placing these selections in the Compatible Mode register. The Mode bit must be set to 1 first to allow the following bits to be changed: DMA1/3, UNI/BIP, SE/DIFF. The default state for all three bits is 0 , which is how the DAS1600 is shipped.

DMA1/3 - $\quad$ When $=1$, DMA channel 3 is selected.
When $=0$, DMA channel 1 is selected.
DMA is allowed only in Compatible Mode, but the DMA channel selection bit can only be changed when the board is in Enhanced mode. The programmed state will be retained when the board is switched back to Compatible mode.

UNI/BIP - $\quad$ When $=1$, Analog Front-End in Unipolar Range
When = 0, Analog Front-End in Bipolar Range
UNI/BIP is used for both Compatible and Enhanced modes, but the bit can only be changed when the board is in Enhanced mode. The programmed state will be retained when the board is switched back to Compatible mode.

SE/DIFF: $\quad$ When $=1$, Analog Front-End in Single-ended Mode
When = 0, Analog Front-End in Differential Mode
SE/DIFF is used for both Compatible and Enhanced modes, but the bit can only be changed when the board is in Enhanced mode. In Compatible mode, the number of channels is limited according to the table below. The programmed state will be retained when the board is switched back to Compatible mode.

MODE: $\quad$ When $=1$, Board is configured in Enhanced Mode, which in general allows for 64 analog input channels, enhanced triggering, FIFO/REP INSW support, etc..
When $=0$, Board is configured in "Compatible" mode, it is software compatible to the DAS1600.

Table 5-7. Comparison of Features: Compatible vs. Enhanced Mode

| Feature | MODE |  |
| :---: | :---: | :---: |
|  | Compatible | Enhanced |
| \# Analog Input Channels | $\begin{aligned} & 16 \text { S-E } \\ & 8 \text { Diff } \end{aligned}$ | $\begin{aligned} & 64 \text { S-E } \\ & 32 \text { Diff } \end{aligned}$ |
| \# Digital I/O | $\begin{aligned} & 4 \text { in } \\ & 4 \text { out } \end{aligned}$ | $\begin{aligned} & 8 \text { in } \\ & 8 \text { out } \end{aligned}$ |
| External Triggering | Combined w/ Pacing on DIN1, gate only | Separate from Pacing, uses DIN1. Can be gate (level) or trigger (edge) |
| External Pacing | Combined w/ Trigger on DIN1 | Separate from Trigger, uses DIN0 |
| DMA | yes | no |
| REP INSW | no | yes |
| Independent Counter at user connector | yes, CTR0 | no, CTR0 used for pre-triggering |
| Burst Mode | Yes | Yes |
| Interrupts | Single ADC DMA | Single ADC <br> REP INSW <br> End of Burst <br> External |
| Pre-trigger | No | Yes |

In Enhanced mode, HC_AI1, HC_AI0 select the interrupt source, which also will determine the acquisition mode the board is to be operated under. These bits are 0 in Compatible mode. Refer to Table 6-8.

Table 5-8. Interrupt Source Coding - Enhanced Mode

| HC_AI1 | HC_AI0 | ADC Acquisition Mode | Interrupt Source |
| :---: | :---: | :---: | :---: |
| 0 | 0 | Single ADC polled | none |
| 0 | 1 | Single ADC interrupt | FIFO Not Empty |
| 1 | 0 | Fifo REP INSW (512 samples) | FIFO $1 / 2$ Full |
| 1 | 1 | Burst REP INSW (2-64 samples) | End of Burst |

If pre-trigger is enabled, (allowed during a REP INSW mode, HC_AI1=1) then an End-of-Acquisition interrupt will be enabled. This interrupt occurs a pre-determined number of conversions after the trigger. Counter 0 is programmed to count conversions after the trigger and will cause the interrupt when it counts down and reaches Terminal Count.

GAIN1, GAIN0 select the front end gain (Table 5-9). They are valid for Enhanced or Compatible Mode.
Table 5-9. Front-End Gain Coding

| GAIN1 | GAIN0 | Analog Input Gain | Analog Input Voltage Range |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 1 | 0 to 10 V or $\pm 10 \mathrm{~V}$ |
| 0 | 1 | 2 | 0 to 5 V or $\pm 5 \mathrm{~V}$ |
| 1 | 0 | 4 | 0 to 2.5 V or $\pm 2.5 \mathrm{~V}$ |
| 1 | 1 | 8 | 0 to 1.25 V or $\pm 1.25 \mathrm{~V}$ |

### 5.10 PACER CLOCK DATA AND CONTROL REGISTERS

## 8254 COUNTER 0 DATA - POST TRIGGER CONVERSION COUNTER

BASE +12 Example, 30Ch, 780 decimal
READ/WRITE

| 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 |

Counter 0 is used to count conversions to stop the acquisition when a known number of samples have occurred. It essentially is gated on when only a 'residual' number of conversions remain. The main counting of samples is done by the Interrupt Service Routine, which will increment each time by 'packets' equal to Fifo $1 / 2$-full (Fifo $1 / 2$-full is 512 for DAS6400). Generally the value loaded into Counter 0 is $\mathrm{N} \bmod 512$, where N is the total count, or the post trigger count, since Total count is not known when pre-trigger is active. Counter 0 will be enabled by use of the ARMED bit (Base+8) when the next-to-last $1 / 2$-full interrupt is processed.
Counter 0 is to operated in Mode 0.

## 8254 COUNTER 1 DATA - PACER DIVIDER LOWER

BASE +13 Example, 30Dh, 781 decimal
READ/WRITE

| 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 |

## 8254 COUNTER 2 DATA - PACER DIVIDER UPPER

BASE +14 Example, 30Eh, 782 decimal

## READ/WRITE

| 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 |

Counter 1 is the lower 16 bits of the 32 -bit pacer clock divider. It's output is fed to the clock input of Counter 2 which is the upper 16-bits of the pacer clock divider. The clock input to Counter 1 is a precision oscillator source, selected by software to be 1 MHz or 10 MHz
Counter 2's output is called the 'Internal Pacer' and can be selected by software to the be the A/D Pacer source. Counters $1 \& 2$ should be configured to operate in 8254 Mode 2.

## 8254 CONTROL REGISTER

BASE +15 Example, 30Fh, 783 decimal WRITE ONLY

| 7 | 6 | 5 | 4 | 2 | 3 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 |

The control register is used to set the operating Modes of 8254 Counters $0,1 \& 2$. A counter is configured by writing the correct Mode information to the Control Register, then the proper count data must be written to the specific Counter Register. The Counters on the 8254 are 16-bit devices. Since the interface to the 8254 is only eight bits wide, Count data is written to the Counter Register as two successive bytes. First the low byte is written, then the high byte. The Control Register is eight bits wide. Further information can be obtained on the 8254 data sheet, available from Intel or Harris.

Every board was fully tested and calibrated before being placed in finished goods inventory at the factory. For normal environments a calibration interval of 6 months to one year is recommended. If frequent variations in temperature or humidity are common then recalibrate at least once every three months. It takes less than 20 minutes to calibrate a CIO-DAS6402.

### 6.1 REQUIRED EQUIPMENT

Ideally, you will need a precision voltage source, a $41 / 2$ digit digital voltmeter ( $51 / 2$ digit for the CIO-DAS-6402/16), a calculator and some wire. If you do not have a precision voltage source, you will need a non-precision source and have to make a few calculations.

You will not need an extender card to calibrate the board but you will need to have the cover off you computer with the power on, so trim pots can be adjusted during calibration. For that reason a plastic screwdriver has been supplied with your CIO-DAS6402. In the event that the screwdriver is dropped into the PC, no damage will result from short circuits.

### 6.2 CALIBRATING THE A/D \& D/A CONVERTERS

The A/D is calibrated by applying a known voltage to an analog input channel and adjusting trim pots for offset and gain. There are four trim pots requiring adjustment to calibrate the analog input section of the CIO-DAS6402. The entire procedure is described in detail in the InstaCal ${ }^{\text {TM }}$, calibration routine.

The CIO-DAS6402 should be calibrated for the range you intend to use it in. When the range is changed, slight variation in Zero and Full Scale may result. These variations can be measured and removed in software if necessary.

## 7 SIGNAL CONDITIONING CIRCUITS

### 7.1 VOLTAGE DIVIDERS

An alternative method of measuring a signal which varies over a range greater than the input range of a digital input, is to use a voltage divider. When correctly designed, it can drop the voltage of the input signal to a safe level the digital input can accept.

Ohm's law states:
Voltage = Current x Resistance
Kirkoff's law states:
The sum of the voltage drops around a circuit will be equal to the voltage drop for the entire circuit.
In a voltage divider, the voltage across one resistor in a series circuit is proportional to the total resistance divided by the one resistor (see formula below).

The object in a voltage divider is to choose two resistors having the proportions of the maximum voltage of the input signal to the maximum allowed input voltage.

The formula for attenuation is:

$$
\begin{array}{ll}
\text { Attenuation }=\frac{\mathrm{R} 1+\mathrm{R} 2}{\mathrm{R} 2} & \begin{array}{l}
\text { For example, if the signal varies between } 0 \text { and } 20 \\
\text { volts and you wish to measure that with an analog } \\
\text { input with a full scale range of } 0 \text { to } 10 \text { volts, the } \\
\text { attenuation (A) is } 2: 1 \text { or just } 2 .
\end{array}
\end{array}
$$

$$
\text { R1 }=(\mathrm{A}-1) \times \mathrm{R} 2 \quad \text { For a given attenuation, pick a resistor and call it }
$$ R2, the use this formula to calculate R1.

Digital inputs often require the use of voltage dividers. For example, if you wish to measure a digital signal that is at 0 volts when off and 24 volts when on, you cannot connect that directly to a digital input. The voltage must be dropped to 5 volts max when on. The attenuation is $24: 5$ or 4.8 .

Using the equation above, if $R 2$ is $1 \mathrm{~K}, \mathrm{R} 1=(4.8-1) \times 1000=3.8 \mathrm{~K}$.
Remember that a TTL input is 'on' when the input voltage is greater than 2.5 volts.

## NOTE

The resistors, R 1 and R 2 , are going to dissipate power in the divider circuit according to the equation $\mathrm{W}=\mathrm{I}^{2} \times \mathrm{R}$; (Current $(\mathrm{I})=$ Voltage/Resistance). The higher the value of the resistance ( $\mathrm{R} 1+\mathrm{R} 2$ ), the less power dissipated by the divider circuit. Here is a simple rule:

For attenuation of $<5: 1$, no resistor should be less than 10 K .
For attenuation of $>5: 1$, no resistor should be less than 1 K .

### 7.2 LOW PASS FILTERS

A low pass filter is placed on the signal wires between a signal and an $A / D$ board. It stops frequencies greater than the cutoff frequency from entering the A/D board's analog or digital inputs.

The cutoff frequency is that frequency above which no variation of voltage with respect to time may enter the circuit. For example, if a low pass filter had a cutoff frequency of 30 Hz , the kind of interference associated with line voltage ( 60 Hz ) would be filtered out but a signal of 25 Hz would be allowed to pass.

In a digital circuit, a low-pass filter is often used to filter an input from a momentary contact button or relay.

A simple low-pass filter may be constructed from one resistor (R) and one capacitor $(\mathrm{C})$. The cutoff frequency is determined according to the formulas below: (Use $\pi=3.14, \mathrm{R}=$ ohms, $\mathrm{C}=$ Farads.)

$$
\mathrm{Fc}=\begin{gathered}
1 \\
2 *---------\mathrm{C}
\end{gathered}
$$

LOW PASS FILTER


Figure 7-2. Low-Pass Filter Theory

### 8.1 CIO-DAS6402/16

Typical for 25 DegC unless otherwise specified.
Power consumption
Icc: Operating
1.17 A typical, 1.67 A max

Analog input section
A/D converter type
Resolution
Programmable ranges
A/D pacing
Data transfer

Polarity
Number of channels

Interrupts

Interrupt enable
Interrupt sources
Trigger sources

A/D Triggering Modes

A/D conversion time
Throughput
Integral Linearity error

No missing codes guaranteed
Gain drift (A/D specs)
Zero drift (A/D specs)
Input leakage current (@25 Deg C)
Input impedance
Absolute maximum input voltage
AD976A, successive-approximation 16 bits
$\pm 10 \mathrm{~V}, \pm 5 \mathrm{~V}, \pm 2.5 \mathrm{~V}, \pm 1.25 \mathrm{~V}, 0$ to $10 \mathrm{~V}, 0$ to $5 \mathrm{~V}, 0$ to $2.5 \mathrm{~V}, 0$ to 1.25 V
Programmable: internal counter or external source (DIN0, rising edge)
Compatible: Byte wide, interrupt, software polled, DMA
Enhanced: Word wide, interrupt, software polled, REP INSW
Above via 1 kilo sample FIFO
Burst Mode (programmable option) @ 100 kHz
Unipolar/Bipolar, software selectable
Compatible: 8 differential or 16 single-ended, software-selectable Enhanced: 32 differential or 64 single-ended, software-selectable

Programmable levels 2, 3, 5, 7, 10, 11, 15
Positive-edge triggered
Programmable interrupt enable for internal and external interrupt End-of-conversion, FIFO not empty, FIFO half full, end-of-burst, external
Compatible: External hardware/software (DIN0)
Enhanced: External trigger/gate (DIN1), edge/level,polarity/edge programmable.
Digital:
Software configurable for Edge (triggered) or level-activated (gated).
Programmable polarity (rising/falling edge trigger, high/low gate).
Pre-trigger:
Unlimited pre- and post-trigger samples. Total \# of samples must be > 512 .
$5 \mu \mathrm{~s}$
100 kHz min
$\pm 2$ LSB max

16 bits
$\pm 7 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$, all ranges
$\pm 2 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$, all ranges
200 nA
Min 10 Meg Ohms
$\pm 15 \mathrm{~V}$

Analog Output
Resolution
Number of channels
D/A type
Voltage Ranges
Offset error
Gain error
Differential nonlinearity
Integral nonlinearity
Monotonicity
Gain drift
Bipolar offset drift
Unipolar offset drift
D/A pacing
D/A trigger modes
Data transfer
Throughput
Settling time ( 20 V step to $\pm 1 / 2 \mathrm{LSB}$ )
Settling time ( 10 V step to $\pm 1 / 2 \mathrm{LSB}$ )
Slew Rate

Current Drive
Output short-circuit duration
Output Coupling
Amp Output Impedance (OP-27)
Miscellaneous

Digital Input / Output
Digital Type
Configuration
Output High
Output Low
Input High
Input Low

16 bits
2 Voltage Outputs
AD660BN
$\pm 2.5, \pm 5, \pm 10,0$ to $2.5,0$ to 5,0 to 10 , switch-selectable
Adjustable to zero by potentiometer
Adjustable to zero by potentiometer
$\pm 1$ LSB max
$\pm 1$ LSB max
Guaranteed monotonic to 15 bits over temperature
$\pm 15 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max
$\pm 5 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max
$\pm 3 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max
Software
N/A
Byte or word update
System-dependent (software-paced)
$12 \mu \mathrm{styp}, 19$ us max
$6 \mu$ styp, 9 us max
2.8 V/uS Typical
$\pm 5 \mathrm{~mA}$ min
40 mA min Continuous
DC
0.1 Ohms max

Double buffered output latches
Update DACs individually or simultaneously (jumper selectable)
Power up and reset, all DAC's cleared to 0 volts (switch selects bipolar or unipolar zero)

Output: 74LS244
Input: 74LS273
Compatible: Two dedicated ports, 4 inputs and 4 outputs
Enhanced: Two dedicated ports, 8 inputs and 8 outputs
2.7 volts @ - 0.4 mA min
0.4 volts @ 8 mA min
2.0 volts min, 7 volts absolute max
0.8 volts max, -0.5 volts absolute min

Counter section
Counter type 82C54
Configuration 3 down-counters, 16 bits each
Counter 0
Compatible mode - Independent Source: Programmable external (CTR0 CLK) or 100 kHz internal source).
Gate: Available at connector (DIN2).
Output: Available at connector (CTR0 OUT).
Enhanced mode - ADC residual sample counter Source: ADC Clock.
Gate: Internal use.
Output: End-of-Acquisition interrupt.
Counter 1-ADC Pacer Lower Divider
Source: 10 MHz oscillator
Gate: Tied to Counter 2 gate, programmable source.
Output: Chained to Counter 2 Clock.
Counter 2 - ADC Pacer Upper Divider
Source: Counter 1 Output.
Gate: Tied to Counter 1 gate, programmable source.
Output: ADC Pacer clock, available at user connector.

Clock input frequency
High pulse width (clock input)
Low pulse width (clock input)
Gate width high
Gate width low
Input low voltage
Input high voltage
Output low voltage
Output high voltage
Environmental
Operating temperature range
Storage temperature range
Humidity

10 MHz max
30 ns min
50 ns min
50 ns min
50 ns min
0.8 V max
2.0 V min
0.4 V max
3.0 V min

0 to $70^{\circ} \mathrm{C}$
-40 to $100^{\circ} \mathrm{C}$
0 to $90 \%$ non-condensing

### 8.2 CIO-DAS6402/12

Typical for $25^{\circ} \mathrm{C}$ unless otherwise specified.

Power consumption
Icc: Operating
Analog input section
A/D converter type
Resolution
Programmable ranges
A/D pacing
Data transfer

Polarity
Number of channels

Interrupts
Interrupt enable
Interrupt sources
Trigger sources

A/D Triggering Modes

A/D conversion time
Throughput
Differential Linearity error
Integral Linearity error
No missing codes guaranteed
Gain drift (A/D specs)
Zero drift (A/D specs)
Input leakage current (@25 Deg C)
Input impedance
Absolute maximum input voltage
1.05 A typical, 1.6 A max

ADS7800, Successive Approximation
12 bits
$\pm 10 \mathrm{~V}, \pm 5 \mathrm{~V}, \pm 2.5 \mathrm{~V}, \pm 1.25 \mathrm{~V}, 0$ to $10 \mathrm{~V}, 0$ to $5 \mathrm{~V}, 0$ to $2.5 \mathrm{~V}, 0$ to 1.25 V
Programmable: internal counter or external source (DIN0, rising edge)
Compatible: Byte wide, interrupt, software polled, DMA
Enhanced: Word wide, interrupt, software polled, REP INSW
Above via 1k sample FIFO
Burst Mode (programmable option) @ 250 kHz

Unipolar/Bipolar, software-selectable
Compatible: 8 differential or 16 single-ended, software-selectable
Enhanced: 32 differential or 64 single-ended, software-selectable
Programmable levels 2, 3, 5, 7, 10, 11, 15
Positive-edge triggered
Programmable interrupt enable for internal and external interrupt
End-of-conversion, FIFO not empty, FIFO half full, end-of-burst, external
Compatible: External hardware/software (DIN0)
Enhanced: External trigger/gate (DIN1), edge/level, polarity/edge programmable.
Digital:
Software configurable for Edge (triggered) or level-activated (gated).
Programmable polarity (rising/falling edge trigger, high/low gate).
Pre-trigger:
Unlimited pre- and post-trigger samples. Total \# of samples must be $>512$.

```
3\mus
333 kHz min
\pm.75 LSB
\pm.5 LSB
12 bits
\pm6pm/ }\mp@subsup{}{}{\circ}\textrm{C}\mathrm{ , all ranges
\pm1 ppm/ }\mp@subsup{}{}{\circ}\textrm{C}\mathrm{ , all ranges
200 nA
Min 10Meg Ohms
\pm15V
```

Analog Output:

Resolution
Number of channels
D/A type
Voltage Ranges
Offset error
Gain error
Differential nonlinearity
Integral nonlinearity
Monotonicity
D/A Gain drift
D/A Bipolar offset drift
D/A Unipolar offset drift
D/A pacing
D/A trigger modes
Data transfer
Throughput
Settling time ( 20 V step to $\pm 1 / 2 \mathrm{LSB}$ )
Slew Rate

Current Drive
Output short-circuit duration
Output Coupling
Amp Output Impedance (AD711)
Miscellaneous

12 bits
2 Voltage Output
AD7237
$\pm 10 \mathrm{~V}, \pm 5 \mathrm{~V}, 0$ to $5 \mathrm{~V}, 0$ to 10 V . Software-programmable
Adjustable to zero by potentiometer
Adjustable to zero by potentiometer
$\pm 1$ LSB max
$\pm 1$ LSB max
Guaranteed monotonic over temperature
$\pm 15 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max
$\pm 5 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max
$\pm 3 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ max

Software
N/A
Byte or word update
System dependent (software paced)
$5 \mu \mathrm{styp}, 8 \mu \mathrm{~s}$ max
$4 \mathrm{~V} / \mu \mathrm{s}$ typ
$\pm 2 \mathrm{~mA}$ min
25 mA indefinite
DC
0.1 Ohms max

Double-buffered output latches
Power up and reset, all DAC's cleared to 0 volts

Output: 74LS244
Input: 74LS273
Compatible: Two dedicated ports, 4 inputs and 4 outputs
Enhanced: Two dedicated ports, 8 inputs and 8 outputs
2.7 volts @ - 0.4 mA min
0.4 volts @ 8 mA min
2.0 volts min, 7 volts absolute max
0.8 volts max, -0.5 volts absolute min

Counter type
Configuration

## 82C54

3 down-counters, 16 bits each

## Counter 0

Compatible mode - Independent
Source: Programmable external (CTR0 CLK) or 100 kHz internal source).
Gate: Available at connector (DIN2).
Output: Available at connector (CTR0 OUT).
Enhanced mode - ADC residual sample counter
Source: ADC Clock.
Gate: Internal use.
Output: End-of-Acquisition interrupt.
Counter 1-ADC Pacer Lower Divider
Source: 10 MHz oscillator
Gate: Tied to Counter 2 gate, programmable source.
Output: Chained to Counter 2 Clock.
Counter 2 - ADC Pacer Upper Divider
Source: Counter 1 Output.
Gate: Tied to Counter 1 gate, programmable source.
Output: ADC Pacer clock, available at user connector.

Clock input frequency
High pulse width (clock input)
Low pulse width (clock input)
Gate width high
Gate width low
Input low voltage
Input high voltage
Output low voltage
Output high voltage

## Environmental

Operating temperature range
Storage temperature range
Humidity
$10 \mathrm{MHz} \max$
30 ns min
50 ns min
50 ns min
50 ns min
0.8 V max
2.0 V min
0.4 V max
3.0 V min

For your notes.

## For your notes.

## EC Declaration of Conformity

We, Measurement Computing Corp., Inc., declare under sole responsibility that the product:
CIO-DAS6402/12
CIO-DAS6402/16

Part Number
12-Bit Analog I/O Board
16-Bit Analog I/O Board
Description
rd

Description
to which this declaration relates, meets the essential requirements, is in conformity with, and CE marking has been applied according to the relevant EC Directives listed below using the relevant section of the following EC standards and other normative documents:

EU EMC Directive 89/336/EEC: Essential requirements relating to electromagnetic compatibility.
EU 55022 Class B: Limits and methods of measurements of radio interference characteristics of information technology equipment.
EN 50082-1: EC generic immunity requirements.
IEC 801-2: Electrostatic discharge requirements for industrial process measurement and control equipment.

IEC 801-3: Radiated electromagnetic field requirements for industrial process measurements and control equipment.
IEC 801-4: Electrically fast transients for industrial process measurement and control equipment.

Carl Haapaoja, Director of Quality Assurance

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